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TITLE

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LA-UR--34-1921

DE64 013907

SUBMITTED TO

16th Power Modulator Symposium, Arlington, VA, June 18-20, 1984.

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THE DEVELOPMENT OF A 3 TESLA - 10 Hz PULSED MAGNET-MODULATOR SYSTEM

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Abstract

In order to support the experimental work one at the Los Alamos Meson Physics Facility new instrumentation and data collection systems of advanced design are developed on a regular basis.

Within the instrumentation system for an experiment at LAMPF, "The Photo-Excitation of the H Ion Resonances," There exists a need for a pulsed air-core electromagnet and modulator system. The magnet must be capable of producing a field strength of O to 3T in a volume of 3.5 cm³. In addition it must be radiation resistant, have a uniform field, operate in a high vacuum with little or no outgassing, and the physical layout of the magnet must provide minimal azmuthal obstruction to both the ion and laser beams.

The modulator must be capable of producing up to a 15KA pulse with duration of two us at a maximum repetition rate of 10 Hz. Modulator layout must be of a low inductance design to improve overall system efficiency. Both magnet and modulator must be extremely reliable so that dara collection time is not lost during the experiment. This paper describes in detail the development of the system.

Introduction

The design criteria for the pulsed magnet system have, in large part, been determined by the physics of the experiment and the optics of both the ion and laser beams used in the experiment. These factors have required the following specifications for the magnet and the modulator.

The magnet must have a maximum field strength of 3T \pm 10% over the volume (3.5cm) of interest, see Fig. 1. In addition it must have good high vacuum properties and little or no outgarsing at full power. While preserving the integrity of the hi-vacuum system the physical device must be able to move vertically \sim 10-12 cm so it may be extracted from the beam line. The device must rotate $^{2}60^{\circ}$ so that 8 (see Fig. 1) may be varied thru its full range. This implies that the magnet must provide minimal azumuthal obstruction to both the ion and laser beams.

Magnet Design

An excensive survey of magnet geometries and related construction difficulties was weighed against the design parameters given. This exercise led us to the Helmioltz configuration shown in Fig. 1. The gap (G) was set at 8mm, allowing ample clearance for both ion and laser beams. If one places the current feet

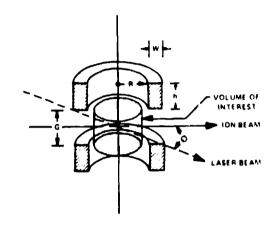


Figure 1. Magnet goometry.

in the shadow of the current return (Fig. 2), θ can be varied thru two 160° sections. By rotating the the assembly 90° , the remaining 40° can be accessed. At this point two computer programs were used to optimize R,W,h and the number of turns for 1T to 3T configurations. Figure 3 shows a cross section of the windings and the dimensions for a 1.0T configuration. The winding materials for the first prototype are shown in Fig. 4. The insulation is reconstructed mica, two layers 0.040mm thick. The conductor is a Sam wide, 0.127mm thick, strip of copper. Two strips of reconstructed mica shoulder the copper, each three layers thick. The coil was wound on an alumina form and an alumina barrel was slipped over the assembly to hold the windings in grace (see Fig. 2). This version of the magnet was found to be unreliable due to breakdown of the mica, leading to arcing between turns.

An improved design, Fig. 5, using Kapton-encapsulated copper windings potted in thermally conducting epoxy was found to be adequate for 1.5T fields. This magnet was then tested with a small SCR Pulser. A field map and a calibration factor were generated. From the field map, Fig. 6, we see the effect of 1/4 extra turn on one coil. This skewed field has since been corrected. The calibration factor for this configuration is 0.98 G/A.



Figure 2. Prototype magnet.

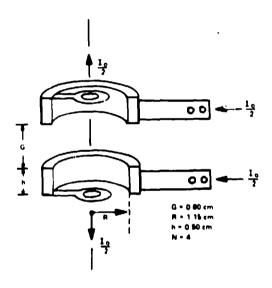


Figure 3. Cross section view of magnet layout.

The inductance of the coil pair is 420nH and the DC coil resistance is $4.6\,\times\,10^{-3}~\Omega_{\odot}$

The Modulator

Figure 7 is a circuit diagram of the modulator. In essence the system is a simple RLC circut with Z_0 set such that 15KA peak current could be achieved within the anode limits of the switch (HY-5). This put $Z_0=.532\ \Omega$ because the magnet inductance (L1,L2) = 420nH. This put the capacitor (c $_1$ -c $_6$ value at $-1.48\mu f$). Figure 8 shows a cutaway viqw of the modulator. This configuration was used so that loop inductance could be minimized by removing the magnet and replacing it with a direct short and then fitting

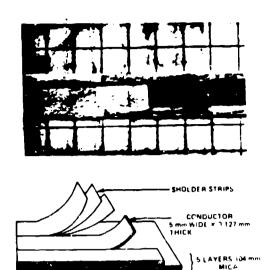


Figure 4. Prototype magnet winding materials.



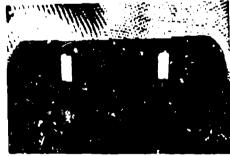


Figure 5. Cross section view of new magnet and winding material.

the current pulse to a theoretical curve. The success of the design was shown by the measured total loop inductance of 64nH. $R_1\!-\!R_6$ are 4.54 Ω such for a parallel combination of .76 Ω which provides a somewhat less then critical damped pulse. The thyratron and the load resistors are water cooled so that the modulator can easily run at the loke average power level. The 0.001 Ω Current Viewing Resistor (CVR) is used to monitor peak current and therefore the atrength of the B field. Power to the magnet is supplied by coaxial line hi-vacuum feed-through. The modulator was then fitted with the magnet and the rystem was tested in high vacuum.

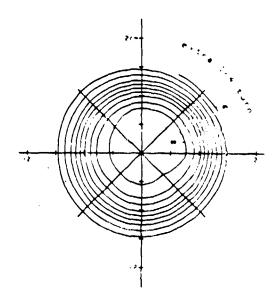


Figure 6. B Field map.

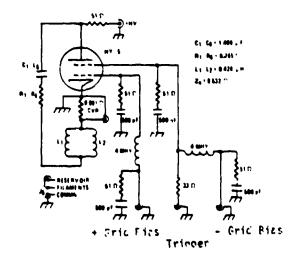


Figure 7. Modulator circuit diagram.

Testing

Figure 9 shows the test setup for the magnet and the modulator. Migh voltage, MV supply current, pulse current and the B field were monitored.

Early on, weak links in the system appeared and in all cases they were related to either coil design or implementation. The modulator has run in an almost flawless manner from the time power was first applied. At this point the modulator has about 3.6 \times 10' pulses, has required wirtually no maintenance, and has caused no down time. Figure 10 shows a typical turrent pulse. The 102 to 90% rise time is 500nS and the FWHM is 1.6µS.

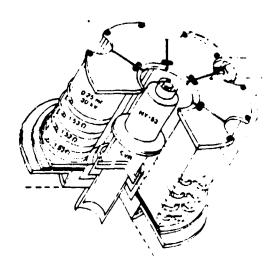


Figure 8. Cross section view of the modulator.

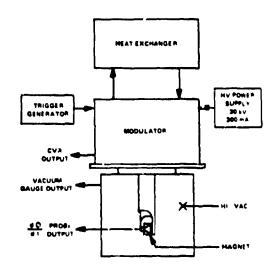


Figure 9. Test setup for magnet and modulator.

The magnet has been a problem that still limits the over all system performance. The problems have been manifold and difficult to solve. The attractive force between the two coils has caused the copper windings to cut the insulation. IR heating and inadequate cooling have caused magnets to run red hot. Local arcing, caused by large L(di/dt) terms, has caused coils to explode.

At this point the problem has been reduced to one of materials research. The insulation used in the magnet windings must have high dielectric etrength and yet be thermally conductive. Figure 5 shows a photo of the new magnet assembly. The coil winding is 0.5mm

oper strip 0.127mm thick bonded between two 0.127mm lck kapton strips. The winding is potted in vacuum a thermally conductive epoxy. Figure 5 also shows cross section of a winding. This construction has owed us to run at peak power of ~800J to iKJ per se, but at these power levels we can only run at a se frequency of ~5HZ, due chiefly to the lack of ling.

Conclusion

Ultimately B fields of >3T and repetition rates HZ will be required. Therefore new cooling systems h greater heat capacities and improved thermal hs are being considered. New coil designs based on ti-layer hybrid technology are being developed. In Ition new thyratrons with higher anode voltage its and higher cathode currents than the HY-5 are in consideration.

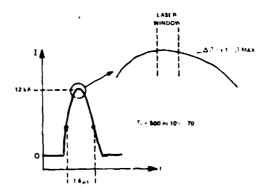


Figure 10. Current pulse.